

STATE ESTIMATION.

Motivation: Man in the woods problem:

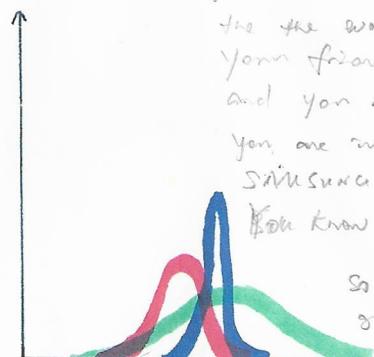
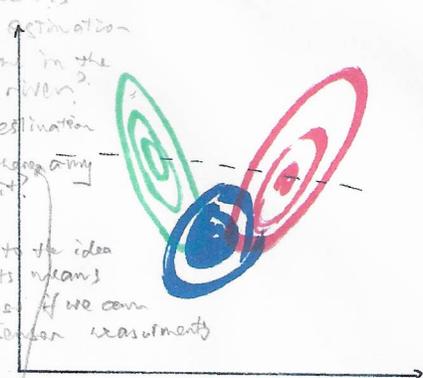
Devise a and b provide independent maps. Each device has Gaussian Noise property. The ellipse describes the region centered on the mean in the distribution that contains 10%, 25% and 50% of data under the distribution. The maximum likelihood estimate of the hidden variable is worked at the center.

Now, a new problem is

What if your estimation tells you you are in the middle of the river?

Obviously the estimation is off. So is there any other way to do it?

This leads us to the idea that more inputs means better result. If we can cooperate the sensor measurements in the past and recursively estimate the state, we can somehow have better results.



Let's start with a simple scenario, A man in the woods problem. Say you and your friend are walking in Hyman Park and you are lost. Your iPhone tells you you are in one location but your friend's iPhone tells you are in another location. You know both of the GPS are off.

So now our motivation is "is there a good way that we can take a trade-off of these two sensors"

Or is there a better way than just take trade-off?

One typical approach is to find the ML of these two sensors by taking into account the quality of / noise of the sensor - cheap -> more noise -> less likely

This is also Biological relevant because in your brain you have the capacity to rely on multiple sensors.

Linear & Nonlinear Process

Linear: In general, A and B can be Non-linear: System, $\dot{x} = Ax + Bu$, which is a first order differential equation. The system is a discrete time system so we can find the solution for that. Finally we can discretize it to be get a time update of the state vector.

$$x_{k+1} = Ax_k + Bu_k + v_k$$

$$z_k = Hx_k + v_k$$

$$v_k \sim \mathcal{N}(0, Q_k)$$

$$w_k \sim \mathcal{N}(0, R_k)$$

$$x_{k+1} = f(x_k, u_k, w_k)$$

$$z_k = h(x_k, v_k)$$

$$v_k \sim \mathcal{N}(0, R_k)$$

Let's follow the idea and start with a slightly more general model to see what we can have.

Complex System Model: Mobile Robot

System model:

Since here you echo, cost, it will give you a quaternion, so here let's use quaternion.

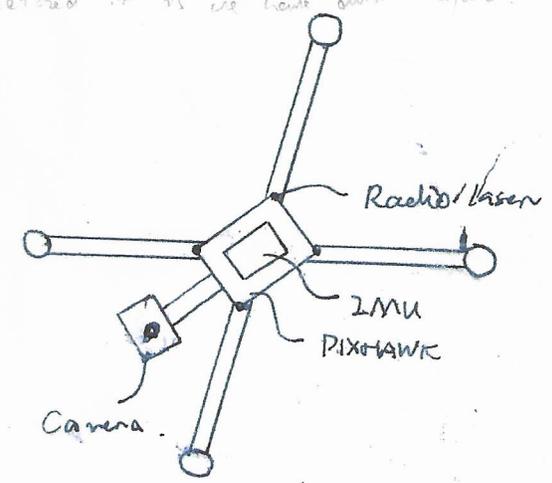
$$x_k = [q_w \ q_x \ q_y \ q_z \ \dot{x} \ \dot{y} \ \dot{z} \ v_x \ v_y \ v_z]$$

The derivation for quaternion:

$$\dot{q}(t) = \frac{w(t)}{2} q(t) =$$

Remember that the derivative of linear system is the change of output is proportional to change of input, so we have $\dot{x} = Ax + Bu$.

now let's find \dot{q} . For quaternion, we have a theorem that $\dot{q}(t) = \frac{w}{2} q(t)$ where w is the angular velocity determined by $q(t)$.



skew symmetric $[0, \omega_x, \omega_y, \omega_z]^T$

$$= \frac{1}{2} \begin{bmatrix} 0 & -\omega_x & -\omega_y & -\omega_z \\ \omega_x & 0 & \omega_z & -\omega_y \\ \omega_y & -\omega_z & 0 & \omega_x \\ \omega_x & \omega_y & -\omega_z & 0 \end{bmatrix} \begin{bmatrix} \dot{\theta}_w \\ \dot{\theta}_x \\ \dot{\theta}_y \\ \dot{\theta}_z \end{bmatrix} = \dot{\omega} \cdot \dot{\theta}$$

The angular velocity can be found from Newton's equations for dynamics.

One thing to take note is that the only thing you measure here is the acceleration and angular rate. So for an IMU, it contains an accelerometer, which measures the acceleration and a gyroscope, which measures angular rate, so these two measurements are in the body frame, but for velocity they are in world frame.

The robot kinematics:

$$[\Delta x, \Delta y, \Delta z]^T = [v_x, v_y, v_z]^T \text{ and position}$$

The derivative of robot velocity is:

$$\begin{bmatrix} \dot{v}_x \\ \dot{v}_y \\ \dot{v}_z \end{bmatrix} = R^{bw} \begin{bmatrix} a_x \\ a_y \\ a_z \end{bmatrix} + \dot{g}$$

$$R^{bw} = \begin{bmatrix} 1 - 2q_x^2 - 2q_y^2 & 2q_x q_y - 2q_z q_w & 2q_x q_z + 2q_y q_w \\ 2q_x q_y + 2q_z q_w & 1 - 2q_x^2 - 2q_z^2 & 2q_y q_z - 2q_x q_w \\ 2q_x q_z - 2q_y q_w & 2q_y q_z + 2q_x q_w & 1 - 2q_x^2 - 2q_y^2 \end{bmatrix}$$

for most of the time you don't give you specific force, so when you are at rest you still have this g in readings, so you need to compensate for this g .

is the rotation matrix of quaternion from world to body.

Then the continuous system model:

$$\dot{x} = [\dot{\theta}, \dot{p}, \dot{v}] = Ax + Bu$$

$$\begin{bmatrix} \dot{\theta} \\ \dot{p} \\ \dot{v} \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} p \\ v \end{bmatrix} = \begin{bmatrix} v \\ 0 \end{bmatrix} + \begin{bmatrix} 0 \\ R_0 + g \end{bmatrix} = \begin{bmatrix} \dot{\omega} & 0 & 0 \\ 0 & 0 & I \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \theta \\ p \\ v \end{bmatrix} + \begin{bmatrix} 0 \\ R_0 + g \\ 0 \end{bmatrix} u$$

So far we have the continuous-time linear system model, and we want to discretize it for implementation purpose. It's one way of digital electronics so the time is discrete.

For implementation, the system equation needs to be discretized by matrix exponential

$$Ad = e^{AT} \text{ For the first diagonal block.}$$

$$e^{A\Delta t} = e^{2 + \omega \sin t + \omega^2 (1 - \cos t)}$$

Use Rodrigues' Formula we have $e^{\frac{\omega \Delta t}{2}} = \begin{bmatrix} \cos(\frac{\omega \Delta t}{2}) & \sin(\frac{\omega \Delta t}{2}) \\ -\sin(\frac{\omega \Delta t}{2}) & \cos(\frac{\omega \Delta t}{2}) \end{bmatrix}$

recall that the solution for 1st order differential equation is (let the initial $f_0 = f(t_k)$)
 $x(t_k) = e^{A(t_k - t_{k-1})} x(t_{k-1}) + \int_{t_{k-1}}^{t_k} e^{A(t_k - \tau)} B u(\tau) d\tau$

$$f_{k+1} = e^{\int_{t_k}^{t_{k+1}} \omega dt} \cdot f_k = A_f \cdot f_k = \begin{bmatrix} r_0 & -r_1 & -r_2 & -r_3 \\ r_1 & r_0 & r_3 & -r_2 \\ r_2 & -r_3 & r_0 & r_1 \\ r_3 & r_2 & -r_1 & r_0 \end{bmatrix} \cdot f_k$$

put into skew-symmetric form

let $t_k - t_{k-1} = \Delta t$, $A(t_k) B(t_k) u(t_k)$ to be constant during the interval where $r_0 = \cos(\frac{\omega \Delta t}{2})$, $r_1 = \frac{\sin(\frac{\omega \Delta t}{2}) \omega_x}{\|\omega\|}$, $r_2 = \frac{\sin(\frac{\omega \Delta t}{2}) \omega_y}{\|\omega\|}$, $r_3 = \frac{\sin(\frac{\omega \Delta t}{2}) \omega_z}{\|\omega\|}$

The second diagonal block

$$\begin{bmatrix} \dot{p} \\ \dot{v} \end{bmatrix} = \begin{bmatrix} 0 & I \\ 0 & 0 \end{bmatrix} \begin{bmatrix} p \\ v \end{bmatrix} + \begin{bmatrix} 0 \\ R_0 + g \end{bmatrix} u$$

Then

$$\begin{bmatrix} p_{k+1} \\ v_{k+1} \end{bmatrix} = A_{pv} \begin{bmatrix} p_k \\ v_k \end{bmatrix} + B_{pv} u_k$$

where

$$A_{pv} = \begin{bmatrix} 1 & 0 & 0 & \Delta t & 0 & 0 \\ 0 & 1 & 0 & 0 & \Delta t & 0 \\ 0 & 0 & 1 & 0 & 0 & \Delta t \\ 0 & 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 \end{bmatrix}, \quad B_{pv} = \begin{bmatrix} T & 0 & 0 & \frac{\Delta t^2}{2} & 0 & 0 \\ 0 & T & 0 & 0 & \frac{\Delta t^2}{2} & 0 \\ 0 & 0 & T & 0 & 0 & \frac{\Delta t^2}{2} \\ 0 & 0 & 0 & T & 0 & 0 \\ 0 & 0 & 0 & 0 & T & 0 \\ 0 & 0 & 0 & 0 & 0 & T \end{bmatrix}$$

$\begin{bmatrix} 6 \times 6 \\ 6 \times 6 \\ 6 \times 6 \\ 3 \times 6 \end{bmatrix}$

$$u_k = [a_x \ a_y \ a_z \ 1]^T$$

In the end, the discretized system equation can be written as

$$\begin{aligned} x_{k+1} &= A_d(k) \cdot \begin{bmatrix} \hat{x}_k \\ [AP_k^T \ v_k^T]^T \end{bmatrix} + B_k u_k \\ &= \begin{bmatrix} A_d & 0 \\ 0 & P_v \end{bmatrix} \begin{bmatrix} \hat{x}_k \\ [AP_k^T \ v_k^T]^T \end{bmatrix} + \begin{bmatrix} 0 \\ B_{pv} \end{bmatrix} u_k \end{aligned}$$

$$\begin{bmatrix} x \\ y \\ z \\ v_x \\ v_y \\ v_z \end{bmatrix} = \begin{bmatrix} x + v_x \Delta t \\ y + v_y \Delta t \\ z + v_z \Delta t \\ v_x \\ v_y \\ v_z \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ (P_{11} + P_{21} + P_{31}) \frac{\Delta x}{2} \Delta t \\ (P_{21} + P_{32} + P_{22}) \frac{\Delta y}{2} \Delta t \\ (P_{31} + P_{32} + P_{33}) \frac{\Delta z}{2} \Delta t \end{bmatrix}$$

Measurement model:

- Laser / radar: 2-Norm model with Gaussian white noise

$$z_i^v = \|[x \ y \ z]^T - [x_i \ y_i \ z_i]^T\| + n_i \sim N(0, \sigma^2)$$

or 2D-range bearing

$$\begin{bmatrix} \|[x \ y]^T - [x_i \ y_i]^T\| \\ \tan^{-1} \left(\frac{y - y_i}{x - x_i} \right) - \theta_i \end{bmatrix}$$

- Camera model:

$$z_c^v = K_c \cdot T_w^c \cdot [x_i \ y_i \ z_i \ 1]^T + n_c \sim N(0, \sigma_c^2)$$

where

$$K_c = \begin{bmatrix} f_x & 0 & c_x \\ 0 & f_y & c_y \\ 0 & 0 & 1 \end{bmatrix}, \quad T_w^c = [R \ t]$$

focal length in terms of pixel, skew coefficient

$$z_c = \begin{bmatrix} u \\ v \\ 1 \end{bmatrix} = \begin{bmatrix} f_x & s & c_x \\ 0 & f_y & c_y \\ 0 & 0 & 1 \end{bmatrix} [R \ T] \begin{bmatrix} x_w \\ y_w \\ z_w \\ 1 \end{bmatrix}$$

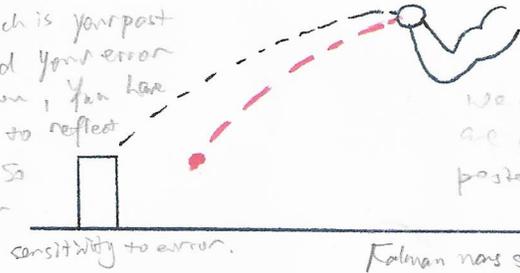


Kalman Filtering: A Frequentist Approach

Let's come back to Kalman filter. So just as we have talked about the variance of each sensor which comes from the sensor noise.

Basic Learning Equation:

What Kalman says is that x_k which is your past is a piece of information, and your error is a second piece of information, you have to combine them by set K to reflect the uncertainty you have. So by knowing this uncertainty you can weigh the two by K .



So the Kalman filter tells us to estimate a parameter in such a way that we minimize our uncertainty about it after we alter our belief about it. So we are going to have some prior belief about what this parameter is. (Location) We are going to make some measurements and we are going to combine these two things to form a posterior belief in a way to minimize our uncertainty.

$$x_k = x_k^p + K_k (z_k - H_k x_k^p)$$

posterior prior sensitivity to error observe predict

$$= (I - K_k H_k) x_k^p + K_k z_k$$

Kalman now saying that when you are trying to do your estimation, you have all the past and you have this current data point. If you talk about Gaussian Noise, all the past are in two parameters, mean and variance. The current estimate can be taken into account the uncertainty of the parameter.

$$z = Hx + R \epsilon \sim N(0, R)$$

Similarly we have the covariance to be:

R is the only variable in z .

R to be the measurement covariance

Kalman says in addition to have these estimates which really are the mean of your distribution. You also have a variance associated with it. So FIND THE MINIMIZE THE VARIANCE

$$\Sigma_k = (I - K_k H_k) \Sigma_k^p (I - K_k H_k)^T + K_k R K_k^T$$

$$= \Sigma_k^p - K_k H_k \Sigma_k^p - \Sigma_k^p H_k^T K_k^T + K_k H_k \Sigma_k^p H_k^T K_k^T + K_k R K_k^T$$

To find the minimum of covariance, use the trace. So in principle Σ_k is a covariance structure in which the diagonal elements are the variance of individual elements.

$$\text{tr}[\Sigma_k] = \text{tr}[\Sigma_k^p] - 2 \text{tr}[K_k H_k^T \Sigma_k^p] + \text{tr}[(H_k \Sigma_k^p H_k^T + R) K_k^T K_k]$$

$$\frac{d}{dK_k} \text{tr}[A X B] = \text{tr}[\Sigma_k^p] - 2 \text{tr}[K_k H_k^T \Sigma_k^p] + (H_k \Sigma_k^p H_k^T + R) \text{tr}[K_k^T K_k]$$

$$\frac{d}{dK_k} \text{tr}[A X B] = \text{tr}[\Sigma_k^p] - 2 H_k \Sigma_k^p H_k^T + (H_k \Sigma_k^p H_k^T + R) K_k^T K_k$$

Take derivative w.r.t R and set it to zero

$$\frac{d \text{tr}[\Sigma_k]}{dK_k} = -2 \Sigma_k^p H_k^T + 2 (H_k \Sigma_k^p H_k^T + R) K_k = 0$$

$$K_k = \Sigma_k^p H_k^T (H_k \Sigma_k^p H_k^T + R)^{-1}$$

So the Kalman gain we are computing is the input H_k and the history of our covariance, so I keep the history of what I have seen, and incorporate finally it into the current estimation.

$$x_k = x_k^p + K_k (z_k - H_k x_k^p)$$

$$\Sigma_k = \Sigma_k^p (I - K_k H_k)$$

the posterior covariance is the prior covariance multiply something less than 1, so your covariance is getting smaller

$$\text{tr}[K K^T] = \text{tr} \begin{bmatrix} K_1^2 & & \\ & K_2^2 & \\ & & \dots & \\ & & & K_n^2 \end{bmatrix}$$

$$K^T K = \begin{bmatrix} K_1 & & \\ & K_2 & \\ & & \dots & \\ & & & K_n \end{bmatrix} \begin{bmatrix} K_1 \\ & K_2 \\ & & \dots \\ & & & K_n \end{bmatrix}$$

Kalman Filtering: A Bayesian Approach

Recall Bayes Theorem:

$$P(x|z) = \frac{P(z|x) P(x)}{P(z)}$$

The posterior probability distribution can be written as:

$$P(x_k | u_{1:k}, z_{1:k}) = \frac{P(z_k | x_k, u_{1:k}, z_{1:k-1}) P(x_k | u_{1:k}, z_{1:k-1})}{P(z_k | u_{1:k}, z_{1:k-1})}$$

Belief posterior:

$$= \eta_k P(z_k | x_k, z_{1:k-1}, u_{1:k}) \cdot P(x_k | z_{1:k-1}, u_{1:k})$$

[current observation doesn't depend on $z_{1:k-1}, u_{1:k}$]

$$= \eta_k P(z_k | x_k) \cdot P(x_k | z_{1:k-1}, u_{1:k}) \quad // \text{Markov Assumption. Given you know the state, missing this observation } z_k \text{ of the world, you can forget what happened at the past}$$

$$= \eta_k P(z_k | x_k) \int_{x_{k-1}} P(x_k | x_{k-1}, z_{1:k-1}, u_{1:k}) \cdot P(x_{k-1} | z_{1:k-1}, u_{1:k}) dx_{k-1}$$

Law of total prob

Current belief depends on control input and last belief, not current observation

Simplify the notation:

This doesn't make sense to have this part

$$= \eta_k P(z_k | x_k) \int_{x_{k-1}} P(x_k | u_k, x_{k-1}) P(x_{k-1} | u_{1:k-1}) dx_{k-1}$$

observation correlation transition prior belief at t-1

$$= \eta_k P(z_k | x_k) \cdot P'(x_k)$$

The prediction for prior belief is also Gaussian.

$$P(x_k | u_k, x_{k-1}) = \mathcal{N}(x_k; A_k x_{k-1} + B_k u_k, Q_k)$$

$$P'(x_k) = \int P(x_k | u_k, x_{k-1}) \cdot P(x_{k-1}) dx_{k-1}$$

$$= \int \exp\left\{-\frac{1}{2} (x_k - A_k x_{k-1} - B_k u_k)^T Q_k^{-1} (x_k - A_k x_{k-1} - B_k u_k)\right\} \cdot \exp\left\{-\frac{1}{2} (x_{k-1} - u_{k-1})^T \Sigma_{k-1}^{-1} (x_{k-1} - u_{k-1})\right\} dx_{k-1}$$

How do I explain this can be gaussian

So the prediction is a New Gaussian with (u's, Σ_k)

$$P'(x_k) = \begin{cases} u_k' = A_k u_{k-1} + B_k u_k \\ \Sigma_k = A_k \Sigma_{k-1} A^T + Q_k \end{cases}$$

$$P(A) = \int_B P(A|B) \cdot P(B)$$

Gaussian is closed under linear operations

The observation model is also Gaussian, so the posterior can be written as:

$$P(x_k) = \eta \cdot P(z_k | x_k) \cdot P'(x_k)$$

$$= \eta \cdot \exp\left\{-\frac{1}{2} (z_k - H_k x_k)^T R_k^{-1} (z_k - H_k x_k)\right\} \cdot \exp\left\{-\frac{1}{2} (x_k - u_k)^T \Sigma_k^{-1} (x_k - u_k)\right\}$$

To find the parameter x_k that maximize the posterior:

$$\arg \max_{x_k} = \arg \max_{x_k} \left\{ \eta \exp\left\{-\frac{1}{2} (z_k - H_k x_k)^T R_k^{-1} (z_k - H_k x_k)\right\} \cdot \exp\left\{-\frac{1}{2} (x_k - u_k)^T \Sigma_k^{-1} (x_k - u_k)\right\} \right\}$$

take the negative sign out, and exponential out.

$$= \arg \min \left\{ (z_k - H_k x_k)^T R_k^{-1} (z_k - H_k x_k) + (x_k - u_k)^T (A_k \Sigma_{k-1} A_k^T + Q_k)^{-1} (x_k - A_k u_k) \right\}$$

Take the derivative w.r.t x_k and set it to zero

$$0 = 2 (H_k^T R_k^{-1} H_k + (A_k \Sigma_{k-1} A_k^T + Q_k)^{-1}) x_k - 2 (H_k^T R_k^{-1} z_k + (A_k \Sigma_{k-1} A_k^T + Q_k)^{-1} A_k u_k)$$

$\Sigma_k^{-1} = \Sigma^{-1}$
if $\Sigma_0 = \infty$, $\Sigma_k^{-1} = \Sigma_k^{-1}$

no prior belief \rightarrow $\Sigma_k = (I - K_k) \Sigma_k$

prior uncertainty $\infty = [\Sigma_k^{-1} - \Sigma_k^{-1} (R_k + H_k \Sigma_k^{-1} H_k^T)^{-1} H_k^T \Sigma_k^{-1}] [H_k^T R_k^{-1} z_k + \Sigma_k^{-1} u_k]$
Woodbury matrix identity

compute the Kalman Gain even if we have no prior belief.

$$= x_k - K_k H_k x_k + [\Sigma_k^{-1} H_k^T R_k^{-1} - \Sigma_k^{-1} H_k^T (R_k + H_k \Sigma_k^{-1} H_k^T)^{-1} H_k \Sigma_k^{-1} H_k R_k^{-1}] z_k$$

$$= x_k - K_k H_k x_k + K_k [H_k \Sigma_k^{-1} H_k^T + R_k] R_k^{-1} - H_k \Sigma_k^{-1} H_k^T R_k^{-1} z_k$$

ML by totally the evidence

$$= x_k - K_k H_k x_k + K_k z_k$$

(Kalman Gain is going to be the ML estimate)

$$= x_k - K_k (z_k - H_k x_k)$$

if I start with Σ_k where often I get my first sample, I end up with CR

$$K_k = \Sigma_k^{-1} H_k^T (R_k + H_k \Sigma_k^{-1} H_k^T)^{-1}$$

And the posterior covariance

$$E[x_k \cdot x_k^T] = E[(x_k + K_k z_k - K_k H_k x_k) (x_k + K_k z_k - K_k H_k x_k)^T]$$

$$= (I - K_k H_k) \Sigma_{k-1} (I - K_k H_k)^T + K_k R_k K_k^T$$

$$= \Sigma_k^{-1} - K_k H_k \Sigma_{k-1}^{-1}$$

Thanks Drs. Leonard & Hager for the support!